

# New rf silicon

Loek Colussi discusses a new bipolar rf transistor technology that rivals GaAs in digital cellular and cordless phone applications.

A new generation of high-performance silicon bipolar rf transistors with transition frequencies in excess of 20GHz has been developed by Philips Semiconductors. These devices are intended for use in low-voltage cordless and cellular telephones.

In addition to small-signal types for use in a phone's rf receiver, these fifth-generation rf wideband transistors include medium-power types. These rival the performance of GaAs devices when used for rf power amplification in a telephone's transmitter. Unlike GaAs devices, however, they operate at high efficiency from a single supply rail – considerably reducing circuit complexity and allowing the design of smaller, lighter, portable phones.

The key to reducing the size and weight of a portable telephone is the use of a smaller battery pack. In order to maximise energy density, this usually means using fewer cells and consequently a lower supply voltage, typically between 3 and 3.6V. Ideally, the entire telephone should then operate from this single supply voltage. Although dc-to-dc converters can be used to create higher supply voltages, they inevitably result in efficiency losses which shorten the telephone's standby and talk times. They also occupy valuable pc board area and increase the telephone's component and assembly costs.

Because the most power-hungry part of a cellular or cordless telephone is its rf power amplifier, it is important that any move to lower supply voltages does not result in a significant loss of efficiency in the power amplifier. The power amplifier should operate at low voltage with high power-added efficiency – i.e. the ratio of rf output power to dc + rf input power. In order to keep component and assembly costs low, it should use as few gain

stages as possible and the minimum number of peripheral components. To reduce test time, it should be alignment-free and provide predictable, reliable, performance.

Figure 1 illustrates how Philips Semiconductors' new wideband rf transistors can be used to meet these objectives in an rf power amplifier for DECT telephones. The design operates from a single 3.6V rail and includes bias circuitry for load power adjustment and on/off switching. In addition, it occupies less than 10 by 20 mm of a standard

two-layer FR4 laminate pc board.

The amplifier delivers 26dBm output power, achieving a power gain of 29 dB and an overall power added efficiency in excess of 50%.

## Amplifying rf power

RF power amplification is achieved using only two of the new wideband devices. Transistor  $Q_1$  – a BFG425W – operates in class-A mode at a  $V_{CE}$  of 3V and a collector current of 30mA. Under these conditions it provides 18dB of gain and an output power level of

Table 1. Measured source and load impedances of the devices used in the DECT power amplifier.

Transistor	Source imp. ( $\Omega$ )	Load imp. ( $\Omega$ )	Conditions
BFG425W	(12+0.7j)	(52+102j)	$V_{CE}=3.0V$ ; $I_C=30mA$ ; $f=1.9GHz$
BFG21W	(9.1-9.5j)	(9.7-6.4j)	$V_{CE}=3.6V$ ; $P_o=26dBm$ ; $f=1.9GHz$

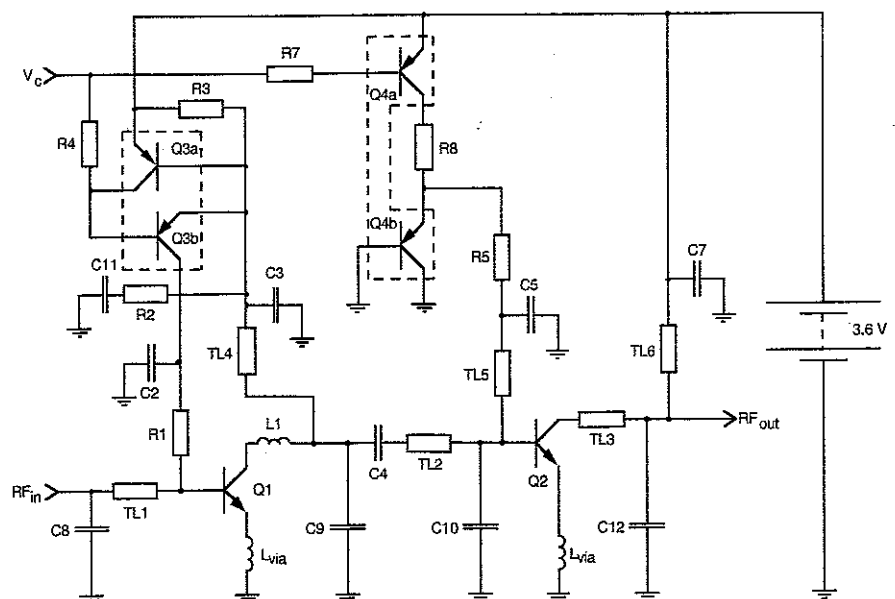


Fig. 1. DECT phone power amplifier using new fifth-generation rf wideband transistors.

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15dBm for -3 dBm rf input.

Transistor  $Q_2$  is a BFG21W medium-power transistor operating in class-AB mode. It drives the telephone's antenna circuit directly. Biased to a base voltage of 0.7V, which results in a quiescent collector current of approximately 1mA, this transistor provides a power gain of 11dB and 26dBm output level.

Under these conditions its collector efficiency is typically 55%.

The measured source and load impedances of the transistors operating under the conditions described above appear in Table 1.

Impedance matching networks are therefore required to provide smooth 50Ω matching throughout the amplifier.

**Impedance matching**

The impedance matching part consists of three separate sections - the input, interstage and output matching networks. Its purpose is to enable the rf transistors to perform optimally with respect to power gain, output power and efficiency.

Fortunately, the inherent impedance levels of the BFG425W and BFG21W as indicated above are not exceptionally high or low, so they are quite easy to match.

At the input, shunt capacitor  $C_8$  and series microstrip line  $TL_1$  match the 50Ω rf source to the base of  $Q_1$ . Base resistor  $R_1$  is used for biasing and has no effect on matching. Between the collector of  $Q_1$  and the base of  $Q_2$ , matching is

done by series inductor  $L_1$ , shunt capacitor  $C_9$  and series transmission line  $TL_2$ .

If pcb area is not critical,  $L_1$  can be replaced by a 3.5mm, 50Ω transmission line. Shunt capacitors  $C_9$  and  $C_{10}$  partly compensate the influence of bias stubs  $TL_4$  and  $TL_5$ , which are both  $<\lambda/4$ .

At the output side of  $Q_1$ , series network  $R_2/C_{11}$  is used to increase the k-factor of the first stage to avoid potential instability below 1GHz. The output match is done by series transmission line  $TL_3$  and shunt capacitor  $C_{12}$ . Again this capacitor also compensates the influence of bias stub  $TL_6$ .

**Biasing**

The biasing part of the power amplifier incorporates a pair of PUMT1 dual p-n-p transistors,  $Q_{3,4}$ . To define the collector current in  $Q_1$ ,  $Q_{3a}$  compares the voltage across  $R_3$  with the forward voltage of its base-emitter junction.

If current in  $R_3$ , i.e.  $Q_1$ 's collector current, increases,  $Q_{3a}$  starts to conduct. This reduces the base drive to  $Q_{3b}$  which in turn reduces the base drive to  $Q_1$ , thereby stabilising  $Q_1$ 's collector current.

For this circuit to work, control voltage  $V_C$  has to be fixed to ground. Voltage on the col-

**Double-poly transistor technology**

Philips Semiconductors' fifth-generation rf wideband transistors are based on a double-polysilicon buried-layer process that yields bipolar transistors with transition frequencies ( $f_T$ ) in excess of 20GHz at low  $V_{CE}$  voltages. Typical power gains of 11dB at 2GHz allow these transistors to be used in the latest generation of digital cordless and cellular telephones - an application previously dominated by GaAs devices.

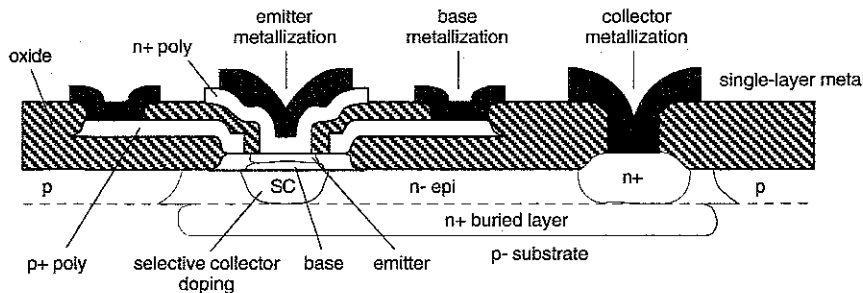
To produce bipolar transistors with cut-off frequencies above 20GHz that will operate at  $V_{CE}$  voltages of 3V or less, it is necessary to achieve base widths in the order of 100 nm. This is accomplished by using the double-polysilicon transistor structure illustrated below, in which deposited polysilicon is used for both the base and emitter connections.

Very steep doping profiles in the base and emitter regions create the very narrow base widths required for a high cut-off frequency, while sub-micron emitter widths of typically 0.5µm - made possible by the self-aligned nature of the process - ensure a high  $f_{max}$ . These sub-micron emitter widths also prevent current crowding effects and help to keep the base resistance low, thereby preventing degradation of power gain. Lateral connections to the base region by p+ polysilicon also help to reduce base

resistance and minimise collector-base (Miller) capacitance.

The n+ layer required for the collector is buried within a p- substrate that is connected to the emitter. This enables the transistor die to be bonded directly to the transistor's emitter lead-outs, reducing its emitter inductance and the thermal resistance of the SOT343R plastic surface-mount package. Large area metallisations for the emitter allow the transistors to handle the high emitter current densities required in medium-power types.

Philips Semiconductors' family of fifth-generation rf wideband transistors currently includes five types. The BFG403W, BFG410W and BFG425W are small signal types optimised for maximum  $f_T$  at collector currents of 3mA, 10mA and 25mA respectively. They feature gains of over 20dB at 2GHz and noise figures as low as 1.2dB. The two medium power transistors in the range are the BFG480W and BFG21W, which have their maximum  $f_T$  values at collector currents of 80mA and 250mA respectively. They provide power gains at a frequency of 2GHz and a  $V_{CE}$  of 3.6V in excess of 14dB and 11dB respectively. Both types achieve typical power added efficiencies greater than 60%.



New wideband bipolar transistors are based on a double-polysilicon buried-layer process yielding transition frequencies ( $f_T$ ) in excess of 20GHz combined with low  $V_{CE}$  voltages.

**Components for the DECT power amplifier.**

$R_1$	560
$R_2$	10
$R_3$	18
$R_4$	100k
$R_5$	10
$R_6$	not required
$R_7$	10k
$R_8$	180
$L_1$	1n8
$C_1$	not required
$C_2$	10n
$C_3$	8p2
$C_4$	8p2
$C_5$	8p2
$C_6$	not required
$C_7$	8p2
$C_8$	1p8
$C_9$	1p8
$C_{10}$	2p7
$C_{11}$	10n
$C_{12}$	2p7
$Q_1$	BFG425W
$Q_2$	BFG21W
$Q_3$	PUMT1
$Q_4$	PUMT1
$TL_1$	Length 6.5mm; Width 0.5mm
$TL_2$	Length 3.0mm; Width 1.2mm
$TL_3$	Length 4.5mm; Width 0.5mm
$TL_4$	Length 7.5mm; Width 0.2mm
$TL_5$	Length 7.5mm; Width 0.2mm
$TL_6$	Length 6.5mm; Width 0.2mm

lector of  $Q_1$  is always 0.6V lower than the supply rail, allowing a 3V collector voltage swing.

The base of class-AB output stage  $Q_2$  is biased by a low impedance voltage source formed by  $Q_{4b}$ . The temperature coefficient of  $Q_{4b}$ 's base-emitter voltage is roughly the same as that for  $Q_2$ , maintaining a quiescent current of approximately 1mA in  $Q_2$ 's collector - despite ambient temperature changes.

Resistor  $R_5$  prevents thermal runaway of  $Q_2$ . Transistor  $Q_{4a}$  is driven by control input  $V_C$  to cut off the base drive to  $Q_2$  during the interval between rf output pulses.

When used in pulsed mode at nominal supply voltage and output level, the load can be mismatched to a voltage/standing-wave ratio less than or equal to 6:1, in all phases, without damage. The power amplifier can also be operated in CW mode provided that 50Ω output matching can be guaranteed under all conditions.

It is possible to increase overall efficiency of the amplifier by a few percent by operating the *BFG425W* in class-AB mode rather than in class-A mode. This also simplifies the biasing circuitry. However, it has the negative effect of reducing the overall power gain, resulting in the need for a higher rf drive level in order to achieve the required 26dBm output power.

If a multi-layer pc board is used, the area occupied by the amplifier can be reduced by burying the transmission lines in the board. Space can be saved by placing the biasing circuitry on the reverse side of the pc board. Inductance of the vias which connect the emitter lead-outs of  $Q_1$  and  $Q_2$  to the ground plane on the reverse of the pc board must be kept lower than 0.1nH in order to maintain rf performance.

**PHS-phone applications**

With only minor modification this power

amplifier is also suitable for use in PHS phones. Although they require a lower rf output power of 21dBm rather than 26dBm, such phones need better linearity performance. To achieve the required linearity, the collector current in the *BFG425W* is reduced to 20mA while the quiescent current through the *BFG21W* is increased to 10mA. In this way, both transistors operate on a more linear part of their gain characteristic.

The *BFG425W*'s collector current can be suitably decreased by increasing the value of  $R_3$  to 22Ω. To increase the quiescent current in the *BFG21W*, the base potential of  $Q_{4b}$  is increased by adding a potential divider between the positive supply rail and ground. A divider comprising 330Ω to ground and 18kΩ to  $V_S$  works well, although other values can be used to achieve an optimum trade-off between linearity and efficiency. ■

**Power gain - even at 2V**

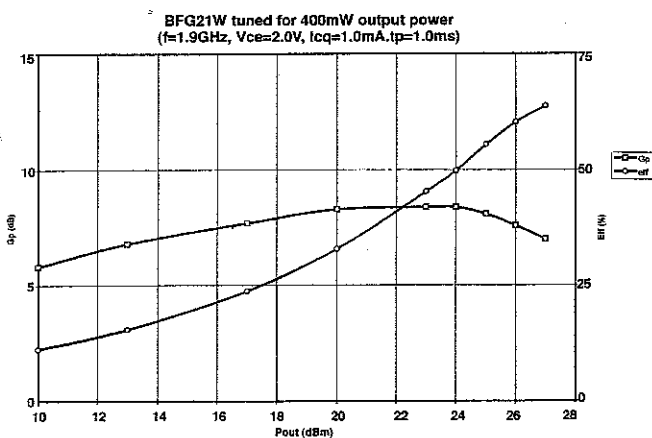
Experiments using a scaled-up version of Philips Semiconductors' *BFG21W* double polysilicon rf transistor indicate that this technology can be used to design DECT power amplifiers that operate from 2.4V battery packs.

As indicated in a) below, the transistors tested provide a power gain,  $G_p$ , of 7dB at the required 27dBm DECT output power level when operating with a  $V_{CE}$  of only 2.0 V. Equally impressive, their power added efficiency at this output level is almost 64%, allowing 2.4V DECT telephones to achieve long standby and talk times.

To compensate for the lower power gain of the output

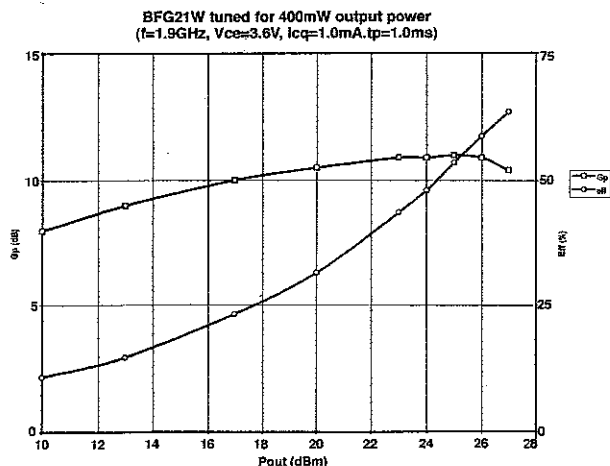
stage at 2.0V (7dB compared to the 10.5dB shown in b) for a  $V_{CE}$  of 3.6V), power amplifiers that use these double-polysilicon transistors would require three gain stages rather than the two described in the main article above. However, the driver and pre-driver stages would each need to provide only 10 to 15dB of power gain in order for the amplifier to be driven at -5dBm.

Detailed test results on the the transistors used in this evaluation can be obtained from Philips Semiconductors' Transistors and Diodes Product Group in Nijmegen, The Netherlands.



**Gain and Efficiency vs.  $P_{out}$**   
 $f=1.9GHz$ ;  $V_S=2.0V$ ;  $I_{CQ}=1mA$ ;  $t_p=1.0ms$ ;  
 tuned for 26dBm output power

$P_{out}$ (dBm)	$P_{in}$ (dBm)	$I_c$ (mA)	$G_p$ (dB)	Eff (%)
10	4.2	45	5.8	11.11111
13	6.2	65	6.8	15.34817
17	9.3	106	7.7	23.64091
20	11.7	152	8.3	32.89474
23	14.6	220	8.4	45.34687
24	15.6	252	8.4	49.83902
25	16.9	285	8.1	55.47856
26	18.4	330	7.6	60.31927
26.5	19.5	350	7	63.81194



**Gain and Efficiency vs.  $P_{out}$**   
 $f=1.9GHz$ ;  $V_S=3.6V$ ;  $I_{CQ}=1mA$ ;  $t_p=1.0ms$ ;  
 tuned for 26dBm output power

$P_{out}$ (dBm)	$P_{in}$ (dBm)	$I_c$ (mA)	$G_p$ (dB)	Eff (%)
10	2	26	8	10.684
13	4	38	9	14.585
17	7	60	10	23.203
20	9.5	88	10.5	31.566
23	12.1	127	10.9	43.641
24	13.1	145	10.9	48.12
25	14	164	11	53.562
26	15.1	188	10.9	58.822
27	16.6	219	10.4	63.57

Thanks for the tip! Jameco has a number of hard to find parts for older computers, even Apple IIs.

As to your question, the trouble with TTL monitors such as the IBM Monochrome Display is that they're completely obsolete—you can get a complete used monitor cheaper than you can get any spare parts, and if you look in the right places, people are giving old monitors away.

The schematic of the IBM Monochrome Display was published in the *IBM Personal Computer Technical Reference* (1982). It's a single page with no explanatory text. Some newer IBM monitor service manuals can be purchased from <http://www.us.pc.ibm.com/cdt/hmm.html>, and Service Editor Sam Goldwasser's excellent notes on monitor repair are online at <http://www.paranoia.com/~filipp>.

Apart from that, there's not much information available. Magazines such as *Computer Hotline* and *Nuts and Volts* used to carry ads for monitor parts and repair guides, but in recent issues the pickings have become very slim. As a last resort, go to the newsgroup [sci.electronics.repair](mailto:sci.electronics.repair) and post your questions; if anyone on earth knows, you'll probably find them there.

## Half A Digit?

I recently started a course in electronics and am puzzled by a term used when describing the features of digital multimeters. Can you explain what is meant by a "4<sup>1</sup>/<sub>2</sub> digit display"? — I. J., Trinidad & Tobago

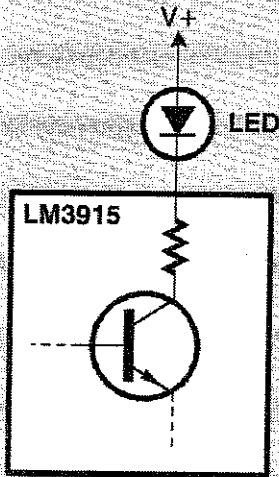


FIG. 4—EACH OF THE LM3915's outputs features an internal switching transistor and a current limiter, which is shown here as a resistor.

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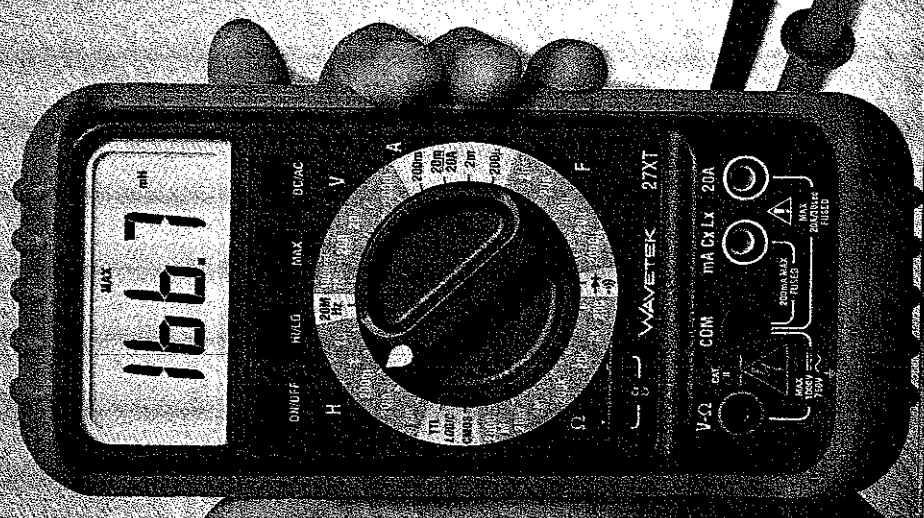
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